

What to do with THz?

Ali M. Niknejad Berkeley Wireless Research Center University of California Berkeley

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Outline

- THz Overview
- Potential THz Applications
- THz Transceivers in Silicon?
- Application 1: THz Radar Transceiver
- Application 2: THz Short Range Communication



• "Transition Region" between Electronics and Photonics $-\lambda=1$ mm-0.1mm (0.3THz $\leq f \leq 3$ THz)

Terahertz Gap : Lack of compact, reliable, tunable source

– THz as Photonics :

Limited by photon energy at THz range (E=hv)

– THz as Electronics :

Limited by the device performance (f_T / f_{max})



Medical Diagnostics[4] Spectroscopy for molecules[5] Remote Gas-sensing[6]

[1] http://eyegillian.wordpress.com/2008/03/10
 [2] Song, MWP2010
 [3] Danylov, THz technology and applications III 2010

[4]http://www.teraview.co.uk/terahertz/[5] Ajito, NTT-technical review 2009[6] Shimizu, NTT-technical review 2009

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THz Transceiver Design Approach in Silicon

Challenges in Silicon Technology

– Active device : Inferior performance (f_T / f_{max}) compared

with III-V compound semiconductors

 Passive device : Large attenuation of THz signal due to high conductive lossy silicon substrate

Advantages:

- Relatively smaller antennas → can realize high antenna
 directivity (gain)
- High bandwidth
- Can integrate antennas on-chip for a true SoC

A "THz" Proof of Concept Radar

Jungdong Park, Shinwon Park, Ali M. Niknejad

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Transceiver Architecture



- FMCW radar transceiver
 - Tx : On-chip antenna + Quadrupler (quadrature push-push)
 - Rx : On-chip antenna + Subharmonic mixer + frequency doubler with an IF buffer for the external measurement

Harmonic Generation with N-Push Structure



$$\dot{i}_T(N\omega) = \sum_{k=1}^N i_k(N\omega) \cos[N(\omega t + \frac{k}{N}2\pi)]$$
$$= N \cdot i(N\omega)$$

- Nth harmonic signals are constructively combined (N·i(Nω)) in current domain while the fundamental signal cancells
- No fundamental signal rejection filter is required
- Desired harmonic element can be optimized with conduction duty cycle (t_o/T)



- Antenna Gain = 6.6 dBi with Radiation Efficiency(η_{rad}) = 44 %
- Each patch is placed in opposite excitation direction for the differential RF input
- GND tap at the center of the patch traps undesired harmonics



- Differential Colpitts VCO + Hybrid + Driving Amplifer
- Differential VCO output power = 3 dBm (Single-ended)
- Hybrid insertion loss = 5 dB
- Driving amplifier gain = 10 dB, $P_{sat} = 6 dBm$ (Differential)



- On-chip antenna + Sub-harmonic mixer + 2nd harmonic LO
- Transformer coupled architecture to provide DC bias and input impedance matching
- Q1, Q2 for emitter degeneneration to reduce switching noise of push-push pairs and acting as ac coupling capacitors at 4fo



- On-chip antenna + Frequency Quadrupler
- Emitter coupled pair for a simplifed matching considering DC path and the low frequency rejection at the collector path

Chip Microphotograph

- Chip fabricated in STM 0.13 μm SiGe process



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Measurement Result (III)

Transceiver Characterization with IF beat signals

- -Tx and Rx fully functional
- -Output frequency is double checked with IF beat frequency



Comparison with reported THz Circuits

Reference	Freq. [THz]	Туре	Output Power [dBm]	NF [dB]	Technology
Huang [ISSCC08]	0.324THz	Quadraple push-push Oscillator	-46 dBm		90nm CMOS
Seok [ISSCC08]	0.41THz	Push-Push Oscillator	-47 dBm	-	45nm CMOS
Öjefors [ISSCC10]	0.65THz	Sub-harmonic Mixer	-	42 dB	0.25µm BiCMOS
Gu [VLSI10]	1.3THz	Quadraple push-push oscillator	Not reported	-	65nm CMOS
Razavi [VLSI10]	0.3THz	Fundamental Oscillator	Not reported	-	65nm CMOS
Öjefors [ISSCC11]	0.82THz	Arrayed Transmitter/Receiver	-17dBm (EIRP 2x2Array)	47dB (53dB)	0.25µm SiGe
Sengupta [ISSCC11]	0.3THz	Arrayed Transmitter	-11 dBm (2x2Array)	-	45nm CMOS
This Work	0.38THz	Single Transceiver	-13 dBm (EIRP)	35dB	0.13µm SiGe

Chip-to-Chip Communication

A "Wireless Bus"

J. Park, S. Kang, S. Thyagarajan, E. Alon, A. Niknejad

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Applications for *very* short range wireless

- If the bandwidth of a wireless bus is sufficiently high, there are many interesting applications for such a technology (chip-to-chip communication)
- Higher frequencies allow higher fractional bandwidths and thus simple modulation schemes can be used to realize high bandwidth links (50 Gbps).
- Higher frequencies (~300 GHz) also allow the on-chip antennas to be smaller than pads, so there's no extra area overhead
- If the power consumption is ~ 0.5W, energy per bit is about 10 pJ/bit, competitive with wired.
- Can it be done?

Future InfoPad Device ...



- Flexible, paper thin, no back-light (natural light only)
- Chips around the periphery communicate wirelessly
- Essentially disposable
- Can "upgrade" device by clipping on another thin layer with more CPU or memory. All connections inside device wireless except DC power !

System Level Design

- In calculation, N (white color) arrayed transceiver is assumed boosts SNR and hence communication range
- For a short range (<2cm), power consumption is comparable between N-OOK and QPSK
- N-OOK is chosen which doesn't require LO synchronization



Challenges

- Path loss at 240 GHz for ~1cm link is around ~36 dB A reliable link requires high Equivalent Isotropically Radiated Power (EIRP)
- Requires design of efficient power amplifiers at mmwave frequencies to enhance EIRP
- Compared to conventional RF design, an LNA cannot be used at 240 GHz as it is beyond the fmax of the device there is no power gain and very high noise figure
- Mixer design at 240 GHz needs to maximize the conversion gain with low noise figure. Elimination of LNA leads to very low available RF signal.
- IF Amplifier following the mixer at 60 GHz needs to provide high gain with high bandwidth to offset the effect of the LNA
- Testing requires design of on-chip PRBS at high data rates (~20 Gbps)
- Distribution of the data stream to the modulator blocks

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